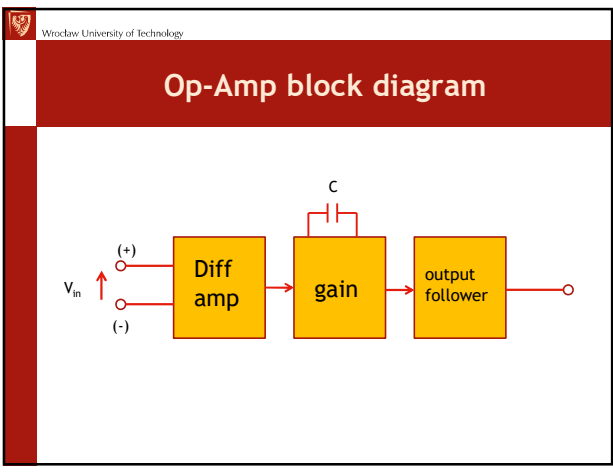
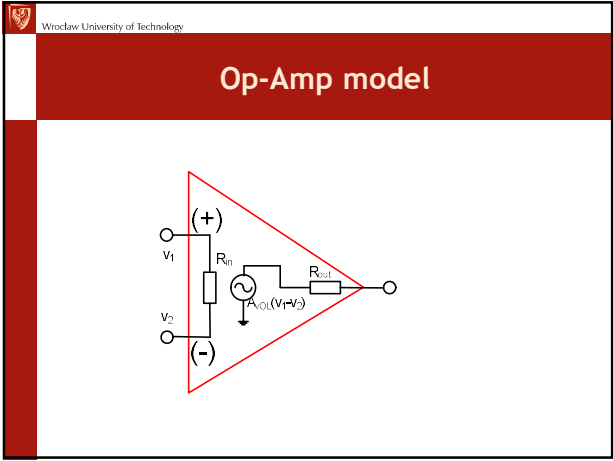
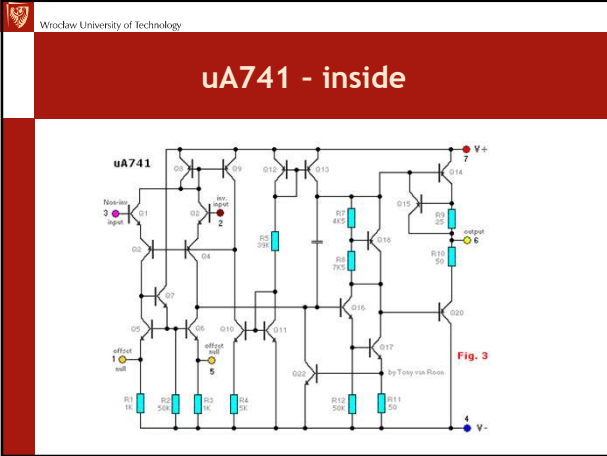


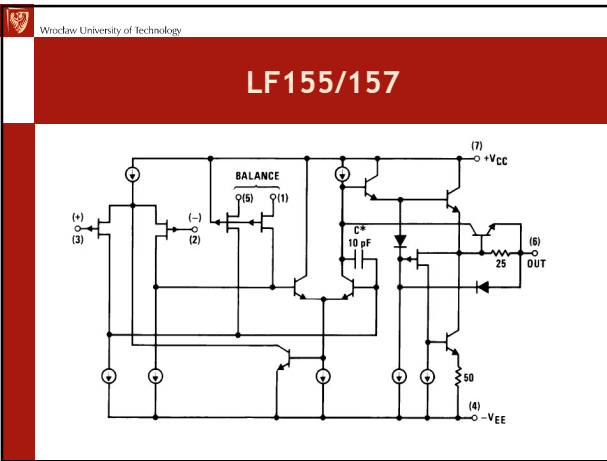
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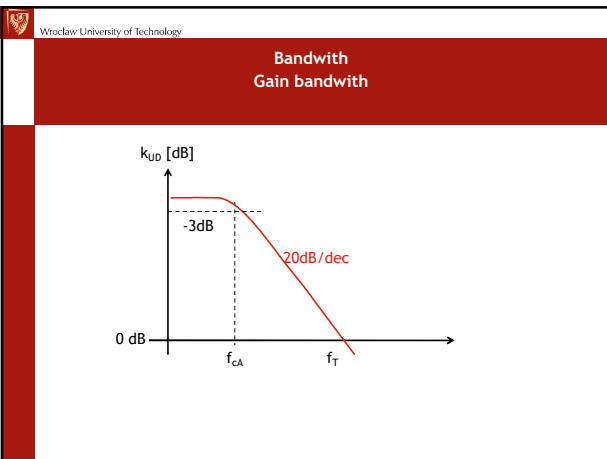
Operational Amplifier Op-Amp











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$$V_{DIFF} = \frac{V_1 - V_2}{2}$$

$$V_{CM} = \frac{V_1 + V_2}{2}$$

$$V_{out} = A_v V_{DIFF} + \frac{A_v}{CMRR} V_{CM}$$

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Common Mode Rejection Ratio

$$V_{out} = A_v V_{DIFF} + \frac{A_v}{CMRR} V_{CM}$$

$$V_{OUT} = \frac{A_v}{CMRR[V/V]} V_{CM}$$

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Input offset voltage

$$U_0 \quad V_{OUT} = 0$$

$$U_0 = 0 \quad V_{OUT} = A_v V_{off}$$

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Voltage offset vs. input bias current

$V_R = I_{B1} R_1$
 $V_R = I_{B2} R_2$

$$V_{out(off)} = (I_B (R + \Delta R) - (I_B + \Delta I_B) R) A_v =$$

$$= (\Delta I_B R + I_B \Delta R) A_v \underset{\Delta R=0}{\approx} \Delta I_B R A_v$$

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Input and output voltage swing

+VCC
 rail-to-rail
 $U_{in(CM)}$
 rail-to-rail
 -VCC

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Slew-Rate Power bandwidth

$\frac{dU}{dt} = 2\pi f U_{max}$
 $\frac{dU}{dt} > SR = \frac{dU_{out}}{dt}$
 $\frac{dU}{dt} < SR = \frac{dU_{out}}{dt}$
 $SR = \frac{dU_{out}}{dt} = 2\pi f U_{out\ max}$
 $\frac{SR}{2\pi U_{out\ max}} = f_p$

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Typical Op-Amp parameters

Parameter	Symbol	Perfect op-amp	uA741	LF157
OL gain	A_{VOL}	infinite	100000	200000
Unity gain frequency	f_T	infinite	1Mhz	20MHz
Input resistance	R_{in}	infinite	2M Ω	10 ¹² Ω
Output resistance	R_{out}	zero	75 Ω	100 Ω
Input bias	$I_{in(bias)}$	zero	80nA	30pA
Input offset current	$I_{in(off)}$	zero	20nA	3pA
Input offset voltage	$V_{in(off)}$	zero	2mV	1mV
CMRR	CMRR	infinite	90dB	100dB

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More Op-Amp parameters

Type	Manufacturer	Offset-voltage	Bias current	Gain-bandwidth product	Slew rate	Operating voltage min/max	OPs per case	Special feature
VVOpAmps: Low cost universal types								
.741	Many	1mV	80nA	1.5MHz	0.6V/ μ s	6/36V	1	high voltage
.324	Many	2mV	45nA	1MHz	0.6V/ μ s	3/32V	4	single supply
AD8034	Analog D.	1mV	1pA	80MHz	80V/ μ s	5/24V	2	fast
AD8098	Analog D.	1mV	0.5pA	325MHz	1000V/ μ s	3/12V	2	damned fast
AD8604	Analog D.	1mV	0.2pA	8MHz	5V/ μ s	3/ 6V	4	RRIO
AD8619	Analog D.	0.5mV	0.2pA	0.4MHz	0.1V/ μ s	2/ 5V	4	RRIO
AD8674	Analog D.	20 μ V	5nA	10MHz	4V/ μ s	8/36V	4	$V_{id} = 3mV/\sqrt{Hz}$
ADA4851	Analog D.	0.6mV	2pA	130MHz	375V/ μ s	3/10V	4	RRO
OP177	Analog D.	20 μ V	1nA	0.8MHz	0.3V/ μ s	6/36V	1	Low offset
LTC6242	Lin. Tech.	50 μ V	1pA	18MHz	10V/ μ s	3/ 6V	4	$V_{id} = 7mV/\sqrt{Hz}$
MAX4094	Maxim	30 μ V	20nA	0.5MHz	0.2V/ μ s	3/ 6V	4	RRIO
MAX4351	Maxim	1mV	8pA	200MHz	490V/ μ s	9/11V	2	RRIO, fast
MAX4405	Maxim	0.3mV	0.2pA	5MHz	3V/ μ s	5/12V	4	RRO
MAX9916	Maxim	0.2mV	1pA	1MHz	0.5V/ μ s	2/ 6V	2	RRIO
LF356	National	1mV	30pA	5MHz	12V/ μ s	8/36V	1	robust
LMC6014	National	1mV	40 fA	1.4MHz	1V/ μ s	5/15V	4	low bias
LMC6484	National	0.1mV	20 fA	1.5MHz	1V/ μ s	3/15V	4	RRIO
LMH6646	National	1mV	0.4pA	55MHz	22V/ μ s	3/12V	2	RRIO
OPA2244	Texas I.	1mV	10nA	0.3MHz	0.1V/ μ s	2/56V	2	$I_C = 40\mu A$
OPA4134	Texas I.	0.5mV	5pA	8MHz	20V/ μ s	5/56V	4	low distortion
OPA4343	Texas I.	2mV	0.2pA	5MHz	6V/ μ s	3/ 6V	4	RRIO
TL084	Texas I.	0.4mV	3pA	10MHz	16V/ μ s	5/16V	4	single supply
TLC274	Texas I.	1mV	0.1pA	2MHz	3V/ μ s	3/16V	4	single supply

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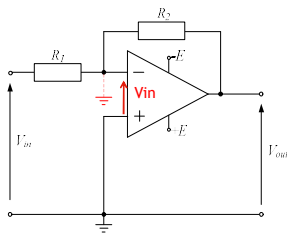
Op-Amp - power supply

The diagram shows an operational amplifier with a differential input. The non-inverting input is labeled 'P' and the inverting input is labeled 'N'. The non-inverting input is connected to a terminal labeled '+E', and the inverting input is connected to a terminal labeled '-E'. The output is labeled 'WY'. The op-amp is powered by two power supplies: 'Power supp. 1' (positive supply) and 'Power supp. 2' (negative supply). The positive supply is connected to the non-inverting input and the negative supply is connected to the inverting input.

Op-Amp common applications

- inverting amp,
- Non-inverting amp
- summing amp
- substracting amp
- integrator
- differentiator

Inverting Amp virtual ground



For a perfekt OA:

$$A_{VOL} = \infty$$

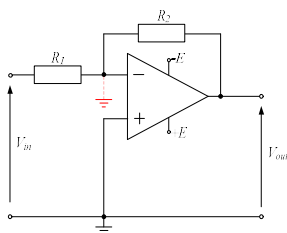
$$r_{in} = \infty$$

so:

$$V_{in} = \frac{V_{out}}{A_{VOL}} = \frac{V_{out}}{\infty} = 0$$

$$I_{in} = \frac{V_{in}}{r_{in}} = \frac{V_{in}}{\infty} = 0$$

Inverting Amp



$$I_{in} = \frac{V_{in}}{R_1} \quad I_{out} = \frac{V_{out}}{R_2}$$

$$I_{in} = -I_{out}$$

$$\frac{V_{in}}{R_1} = -\frac{V_{out}}{R_2}$$

$$A_v = \frac{V_{out}}{V_{in}} = -\frac{R_2}{R_1}$$

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inverting amp (in detail)

$$A_v = -\frac{R_2}{R_1} \frac{1}{1 + \frac{R_1 + R_2}{A_{v,OL} R_1}} \approx -\frac{R_2}{R_1}$$

$$R_m = R_1 \frac{1}{1 + \frac{A_v}{A_{v,OL}}} \approx R_1$$

$$R_{out} = R_{out0} \frac{R_1 + R_2}{R_1(1 + A_{v,OL}) + R_2} \approx R_{out0} \left| \frac{A_v}{A_{v,OL}} \right|$$

$$f_2 \approx \frac{f_T}{A_v}$$

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Bias current compecstion

$$R_d = \frac{R_1 R_2}{R_1 + R_2}$$

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non-inverting amp

$$A_v = \left(1 + \frac{R_2}{R_1}\right) \frac{1}{1 + \frac{R_1 + R_2}{A_{v,OL} R_1}} \approx 1 + \frac{R_2}{R_1}$$

$$R_m = R_{in0}$$

$$R_{out} = \frac{R_{out0}}{A_{v,OL}}$$

$$R_d = \frac{R_1 R_2}{R_1 + R_2}$$

$$f_2 \approx \frac{f_T}{A_v}$$

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Follower

$$A_v = \left(1 + \frac{R_2}{R_1}\right) \frac{1}{1 + \frac{R_1 + R_2}{A_{oL} R_1}} \approx 1 + \frac{R_2}{R_1}$$

$$f_2 = f_T$$

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Frequency characteristic

$$f_2 A_v = f_T \cdot 1$$

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Summing amp

$$V_{out} = -R_2 \sum_{k=1}^N \frac{V_{in k}}{R_{1k}} = - \left(V_{in1} \frac{R_2}{R_{11}} + V_{in2} \frac{R_2}{R_{12}} + V_{in3} \frac{R_2}{R_{13}} + \dots + V_{inN} \frac{R_2}{R_{1N}} \right)$$

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Sumator odwracający

$$V_{out} = -R_2 \sum_{k=1}^N \frac{V_{in_k}}{R_{1k}} = - \left(V_{in1} \frac{R_2}{R_{11}} + V_{in2} \frac{R_2}{R_{12}} + V_{in3} \frac{R_2}{R_{13}} + \dots + V_{inN} \frac{R_2}{R_{1N}} \right)$$

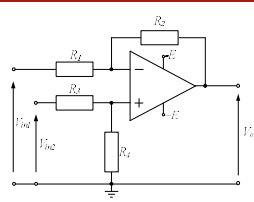
when: $R_{11} = R_{12} = R_{13} = \dots = R_{1N} = R_1$

$$V_{out} = - \frac{R_2}{R_1} (V_{in1} + V_{in2} + V_{in3} + \dots + V_{inN})$$

$$R_d = R_2 \parallel R_{11} \parallel R_{12} \parallel R_{13} \parallel \dots \parallel R_{1N}$$

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Subtractor (DIFFERENCE AMPLIFIER)



$$V_{out} = V_{in2} \left(\frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_4}{R_1} - V_{in1} \frac{R_2}{R_1}$$

when: $R_1 = R_3$ and $R_2 = R_4$

$$V_{out} = \frac{R_2}{R_1} (V_{in2} - V_{in1})$$

$$R_{in1} \approx R_1$$

$$R_{in2} = R_3 + R_4$$

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Subtractor with reference (DIFFERENCE AMPLIFIER with reference)

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Instrumentation Amp

$$V_{out} = \left(1 + 2 \frac{R_2}{R_1}\right) \frac{R_3}{R_4} V_{inDiff}$$

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Instrumentation Amp

$V_{OUT} = (V_{IN2} - V_{IN1}) \left(1 + \frac{2R_5}{R_G} \frac{R_2}{R_1}\right)$
 FOR $R_1 = R_3, R_2 = R_4, R_5 = R_6$

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Instrumentation Amp

offset voltage can be applied

$$V_{out} = (V_{in2} - V_{in1}) \left(1 + \frac{R_1}{R_2} + \frac{2R_1}{R_G}\right)$$

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Integrator

$$I_R = \frac{V_{in}(t)}{R} \quad I_C = C \frac{dV_{out}(t)}{dt}$$

$$I_R + I_C = 0$$

$$V_{out}(t) = -\frac{1}{RC} \int V_{in}(t) dt + U_0$$

$$V_{out} = -\frac{V_{in}}{RC} t + U_0$$

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Differentiator

$$I_C = C \frac{dV_{in}(t)}{dt} \quad I_R = \frac{V_{out}(t)}{R}$$

$$I_R + I_C = 0$$

$$V_{out}(t) = -RC \frac{dV_{in}(t)}{dt}$$

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phase shifter (all pass filter of 1st order)

$$\frac{V_{out}}{V_{in}} = -\frac{1 - sCR_3}{1 + sCR_3}$$

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current - voltage transducer

$$V_{out} = -I_{in}R$$

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Voltage - current transducer

$$I_L = \frac{V_{CC} - V_{in}}{R}$$

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Voltage - current transducer (2)

$$I_L = \frac{V_{in}}{R_1} + \frac{R_2 - R_3}{R_1 R_3} V_{out} \stackrel{R_2=R_3}{=} \frac{V_{in}}{R_1}$$

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INIC

(Current Negative Impedance Converter)

$$I_{in} = \frac{V_{out} - V_{in}}{R_3} = \frac{V_{in} \left(\frac{R_2}{R_1} + 1 \right) - V_{in}}{R_3} = \frac{R_2}{R_3 R_1} V_{in}$$

$$R_{in} = -R_3 \frac{R_2}{R_1}$$

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Gyrator

$Z = \frac{Z_1 Z_3 Z_5}{Z_2 Z_4}$

Gyrator Simulating a 1 H Inductor

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Gyrator

$L = R_1 R_2 C$

$F = 2\pi \sqrt{LC} = 1524 \text{ Hz}$

Signal ω_0 Meter

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Auto-zero Amp ???

TIPS:

- low operatin frequency,
- very low offset

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Summing up

- real and perfect Op-Amp parameters (CMRR, SR, OFFSET, $U_{in}(RR)$, $U_{out}(RR)$, f_T)
- applications:
 - inverting and non-inverting amp,
 - summing amp and subtracting amp,
 - instrumentation amp,
 - integrator and differentiator,
 - idea of an auto zero amp ??
